



# TIME DOMAIN ANALYSIS (A VERY SUPERFICIAL APPROACH)



# Laplace Transform (LT)

- The Laplace Transform (LT) is an integral transform similar to the Fourier Transform
- The LT of a function  $f(t)$  is defined as

$$F(s) = \mathcal{L}\{f\}(s) = \int_0^{\infty} f(t)e^{-st} dt,$$

- $s$  is a *complex variable*
  - This integral does not necessarily exist for all possible  $f(t)$  and  $s$  (If  $s$  has a real part  $>0$ ,  $f(t)$  must not grow faster than  $C e^{\text{Re}(s)t}$ )
- The Inverse Transform is more complicated:

$$f(t) = \mathcal{L}^{-1}\{F(s)\} = \frac{1}{2\pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{st} F(s) ds$$

- where  $\gamma > \text{Re}(\text{all singularities of } f)$ .
- This is a *line integral* in the *complex s-plane*, ‘right’ of all singularities

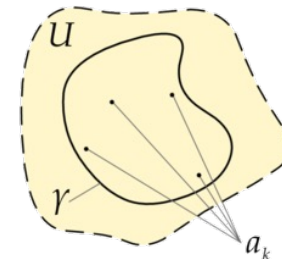


# Reminder (hopefully..): Integration with Residues

This is very simplified!  
 The statements are valid under certain conditions only.  
 Consult a book on Complex Analysis!

- The *Residue Theorem* states that the line integral of a function  $f(z)$  along a *closed* curve  $\gamma$  in the complex  $z$ -plane is  $2\pi i \times$  (the sum of the residues at the *singularities*  $a_k$  of  $f$ ):

$$\oint_{\gamma} f(z) dz = 2\pi i \sum \text{Res}(f, a_k)$$



Wikipedia

- The *residue* is a characteristic of a singularity  $a_k$  (or  $c$  below)
  - For a first order (simple) *pole* at  $c$  (where  $f$  behaves  $\sim$  like  $1/z$ ):

$$\text{Res}(f, c) = \lim_{z \rightarrow c} (z - c) f(z).$$

- More generally, for a pole of order  $n$ :

$$\text{Res}(f, c) = \frac{1}{(n-1)!} \lim_{z \rightarrow c} \frac{d^{n-1}}{dz^{n-1}} ((z-c)^n f(z)).$$



# Example for Integration with Residues

- Assume we want to find  $A = \int_{-\infty}^{\infty} \frac{1}{x^2 + 1} dx$ .
- The function  $f(z) = \frac{1}{z^2 + 1} = \frac{1}{(z + i)(z - i)}$  has poles  $i$  and  $-i$

- The residue at the 'simple pole'  $i$  is:

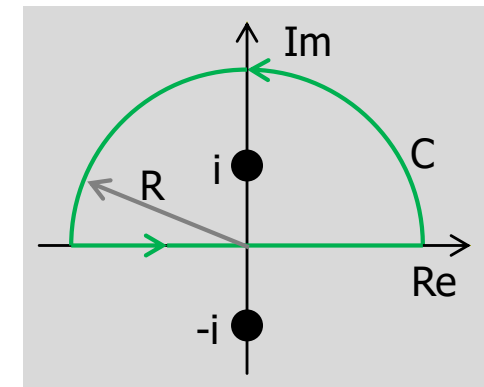
$$Res(f, i) = \lim_{z \rightarrow i} f(z)(z - i) = \lim_{z \rightarrow i} \frac{1}{(z + i)} = \frac{1}{2i}$$

- The line integral along green curve C is

$$\int_C f(z) dz = 2\pi i Res(f, i) = \pi$$

- This is independent of R! With increasing R, the contribution of the upper arc vanishes (the length of the arc rises  $\sim R$ , but  $f$  falls as  $1/R^2$ )

- Therefore  $A = \int_{-\infty}^{\infty} \frac{1}{x^2 + 1} dx = \pi$





## Example 1: LT of 1

- For  $f(t) = 1$ :

$$\mathcal{L}\{1\} = \int_0^{\infty} e^{-st} dt = \left( \frac{-e^{-st}}{s} \right)_0^{\infty} = \frac{-e^{-s\infty}}{s} + \frac{e^{-s0}}{s} = \frac{1}{s}$$

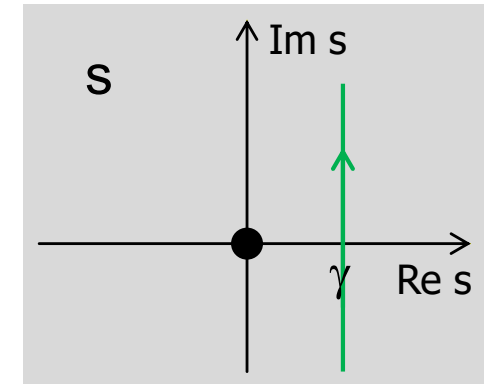
- Valid for  $\text{Re}[s] > 0$  to make sure this vanishes



# Example 1 (inverse): Inverse LT of 1/s

- For  $F(s) = \frac{1}{s}$  :

$$L^{-1} \left\{ \frac{1}{s} \right\} = \frac{1}{2\pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{st} \frac{1}{s} ds$$



- The integrand has just one pole at  $s = 0$ .
- The Residuum is:

$$\text{Res} \left[ e^{st} \frac{1}{s}, 0 \right] = \text{Limit} \left[ (s - 0) e^{st} \frac{1}{s}, s \rightarrow 0 \right] = \text{Limit} \left[ e^{st}, s \rightarrow 0 \right] = 1$$

- So the Integral is

$$\int_{\gamma-i\infty}^{\gamma+i\infty} \dots ds = 2\pi i \text{Res} [ \dots, 0 ] = 2\pi i$$

- And we just have

$$L^{-1} \left\{ \frac{1}{s} \right\} = 1$$

(the arc contribution at infinity vanishes again...)



# Properties of Laplace Transforms

- For  $f(t) = \mathcal{L}^{-1}\{F(s)\}$ ,  $g(t) = \mathcal{L}^{-1}\{G(s)\}$  we have:

Function	Laplace Transform	
$af(t) + bg(t)$	$aF(s) + bG(s)$	Linearity
$f'(t)$	$sF(s) - f(0)$	Derivative
$(f * g)(t) = \int_0^t f(\tau)g(t - \tau) d\tau$	$F(s) \cdot G(s)$	Convolution (Faltung)
$\int_0^t f(\tau) d\tau = (u * f)(t)$	$\frac{1}{s}F(s)$	Integration
$f(t - a)u(t - a)$	$e^{-as}F(s)$	Time Shift

u(t): Step function

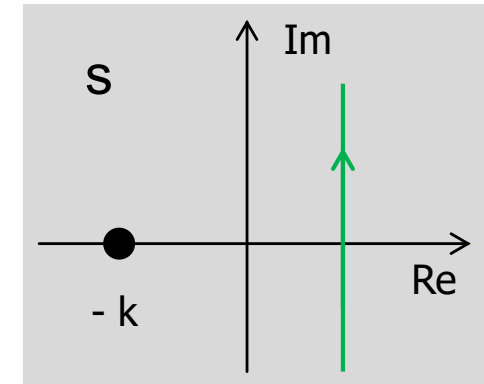


## Example 2 ('frequency shift'):

- For  $F(s) = \frac{1}{s+k}$  :

$$L^{-1}\{F[s]\} = \frac{1}{2\pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{st} \frac{1}{s+k} ds$$

- The pole is now at  $s = -k$ .



- The Residuum is:

$$\text{Res}\left[e^{st} \frac{1}{s+k}, -k\right] = \text{Limit}\left[(s+k) e^{st} \frac{1}{s+k}, s \rightarrow -k\right] = e^{-kt}$$

- And we just have

$$L^{-1}\left\{\frac{1}{s+k}\right\} = e^{-kt}$$





# Why is Laplace Transform so Useful ?

- Differential / Integral equations in  $t$  can be converted to analytical equations in  $s$ , where they can be solved
- EQ( $t$ )  $\rightarrow$  transform to  $H(s)$   $\rightarrow$  Solve in  $s$   $\rightarrow$  Transform back

- Example: Radioactive Decay

- $f[t]$ : Number of atoms at time  $t$

- The # of decaying atoms is prop. to # of atoms:  $\frac{df[t]}{dt} = -\lambda f[t]$

- With  $F[s] = \text{LT}(f[t])$ :  
( $f[0] = N_0$  is initial number of atoms)

$$\text{LT} \curvearrowright s F[s] - f[0] = -\lambda F[s]$$

- This can be solved in  $s$ -domain:

$$F[s] = \frac{N_0}{s + \lambda}$$

- Transforming back (see example) gives:  $f[t] = N_0 e^{-\lambda t}$

$$\text{LT}^{-1} \curvearrowright f[t] = N_0 e^{-\lambda t}$$

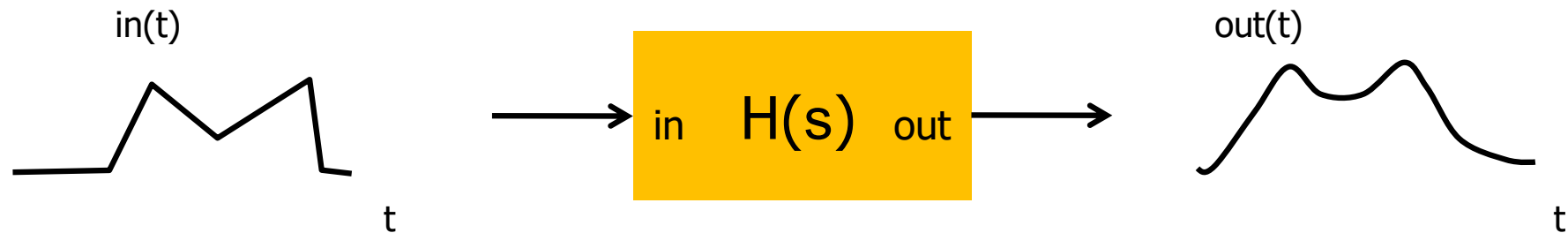


# LAPLACE TRANSFORM AND TRANSFER FUNCTION



# Time Response

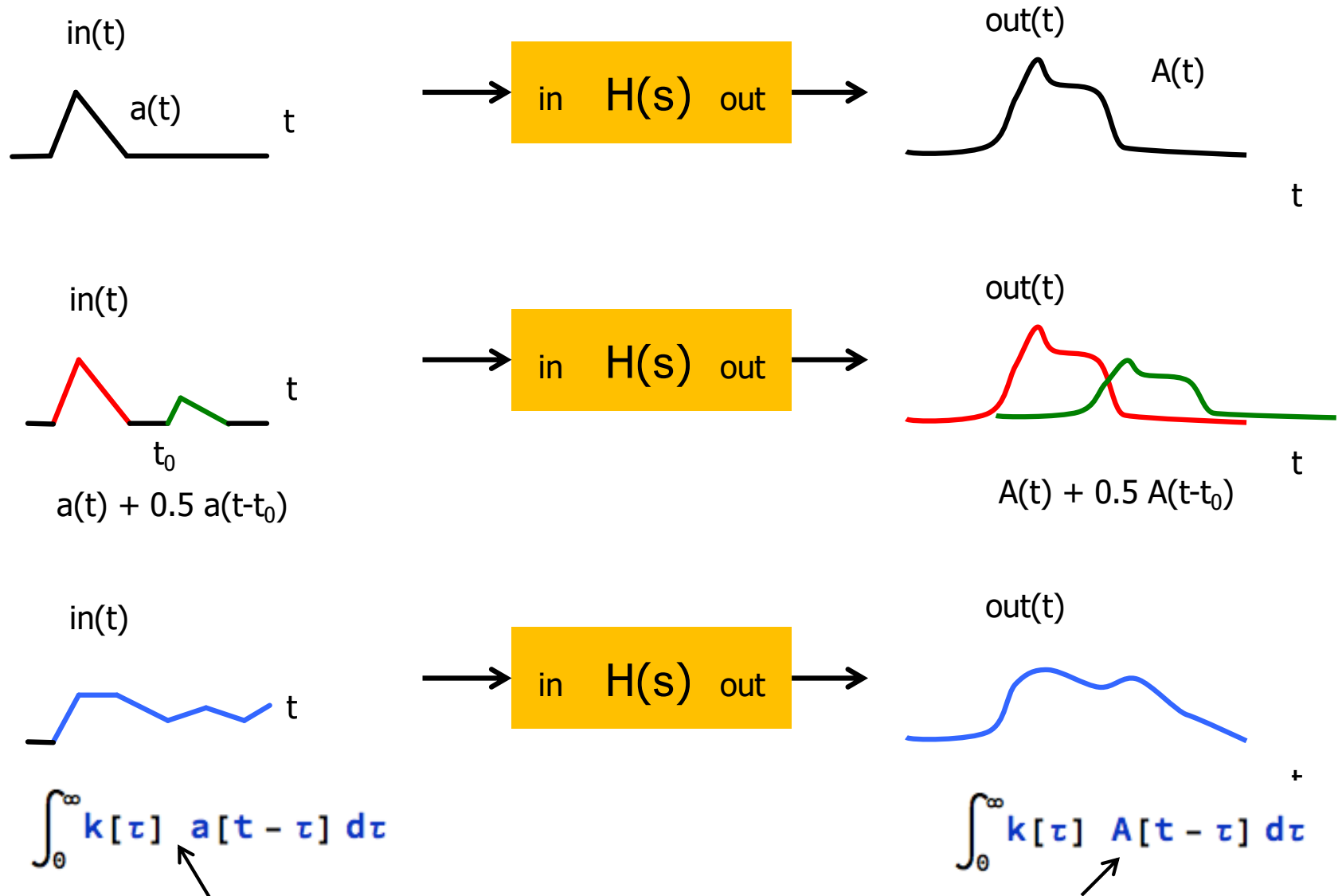
- The **transfer function** tells us how sine inputs are modified by the system, i.e. what happens in the **frequency domain**
- How can we get the **time response** for an arbitrary input?



- For a *linear, time invariant (LTI)* system, we can have:
  - The response of a  $k \times$  larger input pulse is just  $k \times$  larger
  - The response for a time shifted input is time shifted
- For such a system we can
  - express the input signal as a superposition of 'simple' signals
  - Calculate the output for each 'simple' component
  - Superimpose the outputs



# Illustration



Note that the integrals are CONVOLUTIONS ('Faltung') of two functions!



## Clever Choice of the 'nominal input' $a[t]$

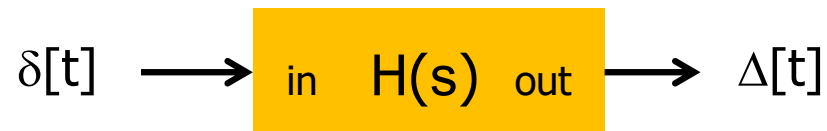
- To make the convolutions as simple as possible, it is best to chose  $a[t]$  to be Dirac Delta 'function'
- For any input function we can write

$$f_{in}[t] = \int_{-\infty}^{\infty} f_{in}[\tau] \delta[t - \tau] d\tau$$

- The output is then just

$$f_{out}[t] = \int_{-\infty}^{\infty} f_{in}[\tau] \Delta[t - \tau] d\tau$$

where  $\Delta[t]$  is the response of the circuit to a  $\delta[t]$  input, the '**delta response**':



Note: I am a bit sloppy here with integration limits..



# What is the Delta Response $\Delta[t]$ ?

- We do not know  $\Delta[t]$ , but: it turns out that its LT is just the transfer function!

The Laplace Transform of the Delta Response of a circuit is just given by its transfer function  $H[s]$

- Knowing that  $LT(\Delta[t]) = H[s]$ , what is  $\Delta[t]$  ?  
It's the Inverse LT:

$$\Delta[t] = LT^{-1} \{H[s]\}$$

- Why is this?
  - If we write down Kirchhoff's rules in the time domain, we get differential / integral equations.
  - The 'topology' of the equations is the same as using complex impedances.
  - If we transform this, we can get the impulse response



# General Time Response

- Start from  $f_{\text{out}}[t] = \int_{-\infty}^{\infty} f_{\text{in}}[\tau] \Delta[t - \tau] d\tau$

- Laplace transform both sides and use Convolution rule:

$$F_{\text{out}}[s] = \text{LT} \left\{ \int_{-\infty}^{\infty} f_{\text{in}}[\tau] \Delta[t - \tau] d\tau \right\} = F_{\text{in}}[s] \text{LT} \{ \Delta[t] \}$$

- Use our knowledge that  $\text{LT} \{ \Delta[t] \} = H[s]$

$$F_{\text{out}}[s] = F_{\text{in}}[s] H[s]$$

- And transform back:

$$f_{\text{out}}[t] = \text{LT}^{-1} \{ \text{LT} \{ f_{\text{in}}[t] \} H[s] \}$$

To calculate the time response of a circuit to an **arbitrary** input  $f[t]$ :

1. Laplace Transform  $f[t]$ , yielding  $F[s]$
2. Multiply with the Transfer function  $H[s]$
3. Laplace Transform back



# Important Input Functions

- The most important input to test a circuit is the Unit step:
  - It is often called  $u[t]$ , Heaviside Step function, UnitStep,...

$$u(t) \xrightarrow{\text{LT}} \frac{1}{s}$$

- For a Shifted Step, use Time Shift rule:

$$u(t) \xrightarrow{\text{LT}} \frac{e^{-s\tau}}{s}$$

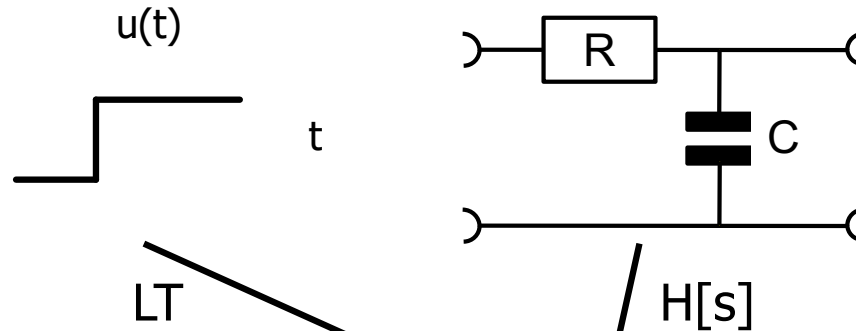
- A rectangular Pulse is just the difference of two Unit Steps
- For very short input signals (charge deposition in detector), input is the Dirac Delta, with  $LT = 1$ .





# Example 1: Response of Low Pass to Step Input

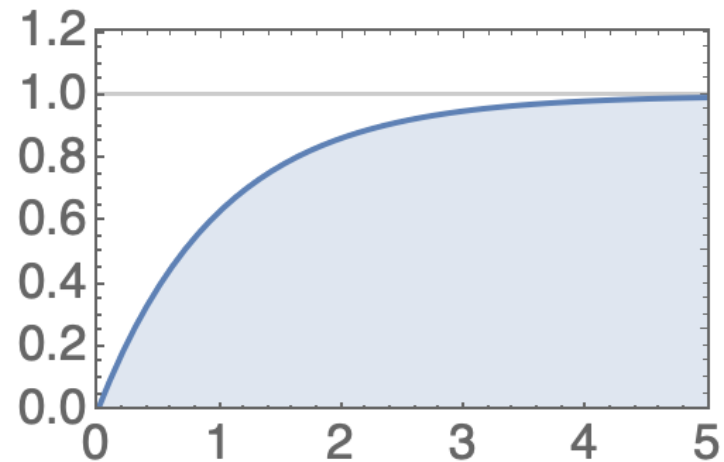
- Consider



LT

InverseLaplaceTransform  $\left[ \frac{1}{s} \frac{1}{1 + s\tau}, s, t \right]$

$$1 - e^{-\frac{t}{\tau}}$$



- This is the *step response*

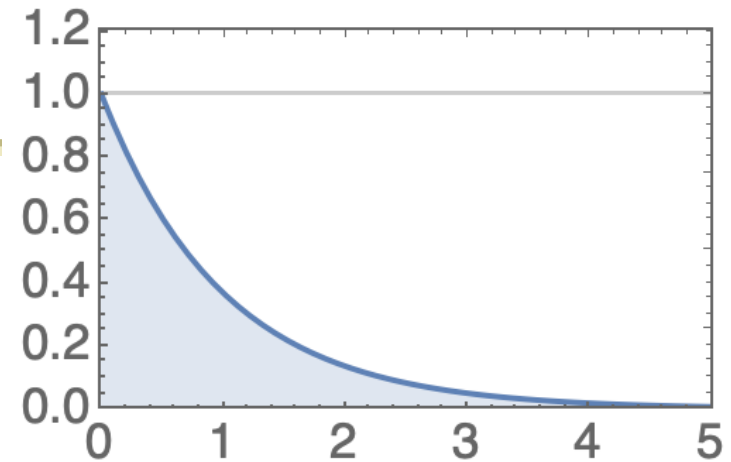


# Watch out! Do not forget 1/s!

- What is  $\text{LT}^{-1}[\text{H}[s]]$  ???, e.g.

$$\text{InverseLaplaceTransform}\left[\frac{1}{1 + s \tau}, s, t\right]$$

$$\frac{e^{-\frac{t}{\tau}}}{\tau}$$

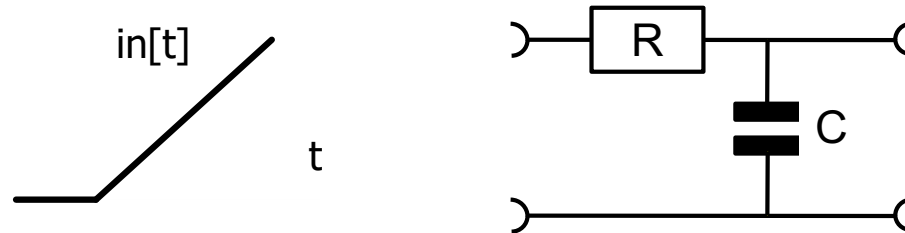


- That is the output for a **Dirac pulse** at the input (with  $\text{LT}[\delta[t]] = 1$ ), i.e. that is the *impulse response*
- See page 14...



# Example 2: Response of Low Pass to Slope

- Now Consider a linear input ramp  $in[t] = k t$



```
In[301]:= IN[s_] = LaplaceTransform[k t UnitStep[t], t, s]
```

- The LT is

$$Out[301]= \frac{k}{s^2}$$

- So our response is

```
= f[t_] = InverseLaplaceTransform[IN[s] H[s], s, t]
```

$$k \left( t - \tau + e^{-\frac{t}{\tau}} \tau \right)$$

```
▼ In[46]:= Limit[D[f[t], t], t -> ∞]
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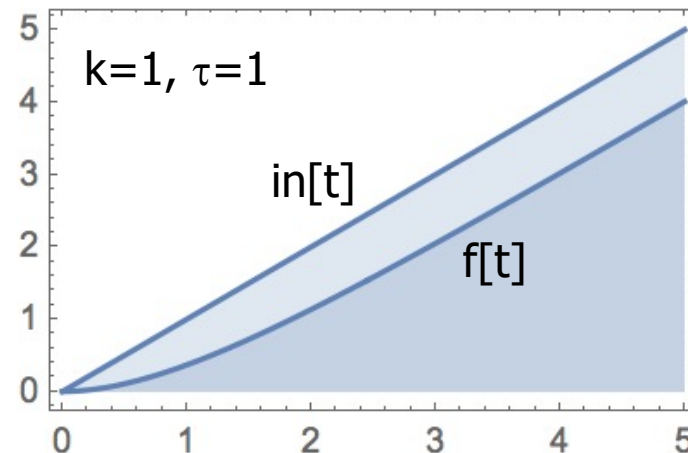
Out[46]=

$$k$$

```
▼ In[47]:= Limit[f[t] - k t, t -> ∞]
```

Out[47]=

$$-k \tau$$

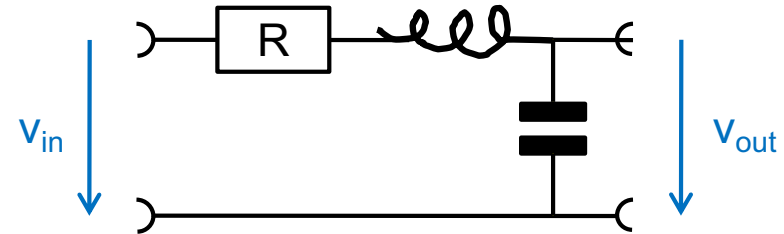




# Example 3: Step Response of RLC

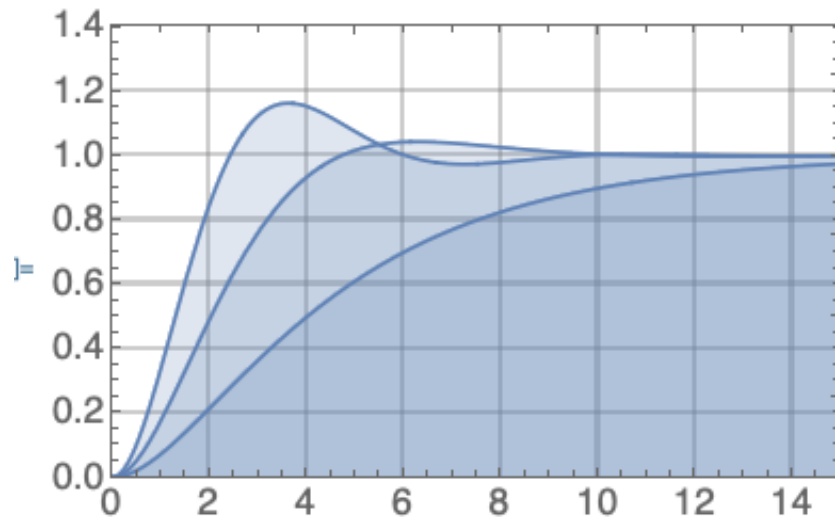
- Consider the RLC circuit

$$H[s] = \frac{1}{1 + CRs + CLs^2}$$



$$f[t_] = \text{InverseLaplaceTransform}\left[\frac{HH}{s}, s, t\right] // \text{FullSimplify}$$

$$1 - e^{-\frac{Rt}{2L}} \text{Cosh}\left[\frac{\sqrt{-\frac{4L}{C} + R^2} t}{2L}\right] - \frac{e^{-\frac{Rt}{2L}} R \text{Sinh}\left[\frac{\sqrt{-\frac{4L}{C} + R^2} t}{2L}\right]}{\sqrt{-\frac{4L}{C} + R^2}}$$



$$R=1, L=1, C=1,2,5$$